TITLE OF THE INVENTION

SPREAD-SPECTRUM COMMUNICATION METHOD AND APPARATUS

5   ## BACKGROUND OF THE INVENTION

Field of the Invention

The present invention relates to a spread-spectrum communication method and apparatus.

10   ### Description of the Related Art

In a time-sharing communication method for converting data to bursts, in order to receive and demodulate data bursts it is necessary to establish synchronization with the data bursts to be received. It is also necessary that only

15   desired information addressed to the receiving end be detected and extracted from the received signals. In addition, conflict between one station and another station must be avoided. Accordingly, in general, in such a communication method data is transmitted in accordance with

20   a regular format.

Fig. 1 shows an example of a format for a data burst in the above communication method. For example, data bursts each include a preamble consisting of a synchronization code (SY), a unique word (UW) and a station-identification code

25   (ID), and data (DA). Between the data bursts there is a

guard time (GT).

A receiving end which receives the data bursts uses the synchronization code in the preamble period to perform reproduction of a carrier, input of automatic gain control

5 (AGC), establishment of clock synchronization, and so forth. The receiving end further detects the unique word (UW) and the station-identification code (ID), and when it perceives that the successive data (DA) is desired data addressed to itself, it holds a reproduced carrier, AGC, clock

10 synchronization and so forth until the data terminates, and it demodulates the data.

However, this communication method causes an error in the reference clock frequency between the transmitting and receiving ends. Thus, with the lapse of time, the receiving

15 end's clock which has held the established synchronization in the preamble, also has increased synchronization errors with respect to the transmitting end's clock. In addition, for example, if the transmission line is wireless, communication quality may vary with time, which results in

20 the possibility of the input AGC held in the preamble losing its optimum condition with the lapse of time. According to this communication method, the maximum time during which data can be transmitted with one data burst is limited by the time during which synchronization precision, AGC

25 precision and so forth can be maintained.

According to the above communication method, when a large amount of digital data (e.g., image data or the like) adapted for recent multimedia applications is transmitted, as shown in Fig. 2, data is divided into a plurality of

5 portions, and the complete data must be transmitted as a plurality of data bursts. As a result, the preamble time and the guard time with respect to the time used for transmission of the actual data increases, and there is the possibility of an interrupting burst from another station

10 occurring between the divided data bursts. Consequently, the data throughput deteriorates.

In order to extend the maximum time during which the data can be transmitted with one data burst, a precise frequency oscillator, and a complicated synchronization

15 circuit or AGC must be used, which disadvantageously requires an expensive, large-sized apparatus.

SUMMARY OF THE INVENTION

Accordingly, it is an object of the present invention

20 to provide a spread-spectrum communication method and apparatus having a high throughput.

It is another object of the present invention to provide a spread-spectrum communication method and apparatus adapted for performing large-amount data communication.

25 It is a further object of the present invention to

provide a spread-spectrum communication method and apparatus in which an adjustment period for receiving spread spectrum data is provided in a period for transmitting spread spectrum data.

5   It is a still further object of the present invention to provide a spread-spectrum communication method and apparatus which communicate an adjustment signal for adjusting reception of spread spectrum data in a plurality of divided data-communication periods.

10   Other objects of the present invention will be apparent from the embodiments described below, based on the attached drawings.

To this end, according to a first aspect of the present invention, the foregoing objects have been achieved through

15   provision of a spread spectrum communication method comprising the steps of: dividing a communication period for spread spectrum data into a plurality of communication periods; and providing an adjustment period for receiving the spread spectrum data between one data-communication

20   period and another data-communication period.

The spread spectrum communication method may further comprise the step of synchronizing a spread code in the adjustment period.

The spread spectrum communication method may further

25   comprise the step of providing the adjustment period prior

to the plurality of data-communication periods.

The spread spectrum communication method may further comprise the step of holding the adjusted setting of the receiving end in the data-communication period.

The spread spectrum communication method may further comprise the step of holding gain in the data-communication period.

The spread spectrum communication method may further comprise the step of communicating code-division-multiplexed data in the data-communication period.

The spectrum communication method further comprising the step of providing the adjustment period prior to the plurality of data-communication periods, may still further comprise the steps of establishing the setting of a receiving end in the adjustment period prior to the plurality of data communication periods; and correcting the established setting in the adjustment period between one data-communication period and the next data-communication period.

Preferably, in the spread spectrum communication method further comprising the step of providing the adjustment period prior to the plurality of data-communication periods, the gain for the adjustment in the adjustment period prior to the plurality of data-communication periods is larger than the gain for the adjustment in the adjustment period

between the one data-communication period and the next data-communication period.

Preferably, in the spread spectrum communication method further comprising the step of providing the adjustment period prior to the plurality of data-communication periods, an adjusting signal communicated in the adjustment period prior to the plurality of data-communication periods is longer than an adjusting signal communicated in the adjustment period between the one data-communication period and the next data-communication period.

The spread spectrum communication method further comprising the step of communicating code-division-multiplexed data in the data-communication period may further comprise the step of communicating a signal not multiplexed by code division multiplexing, in the adjustment period.

According to a second aspect of the present invention, the foregoing objects have been achieved through provision of a spread spectrum communication apparatus comprising: data communication means for communicating spread spectrum data in a plurality of divided data-communication periods; and adjustment-signal communication means for communicating an adjustment signal for adjusting reception of spread spectrum data between one data-communication period and another communication period.

The spread spectrum communication apparatus may further comprise adjustment means for establishing the setting of a receiving end in accordance with the adjustment signal prior to the plurality of data-communication periods and

5 correcting the established setting in accordance with the adjustment means between the one data-communication period and the next data-communication period.

The spread spectrum communication apparatus may further comprise holding means for holding the setting of the

10 receiving end in the data-communication period.

BRIEF DESCRIPTION OF THE DRAWINGS

Fig. 1 is a chart showing details of the format of a data burst according to the related art.

15 Fig. 2 is a chart showing a case where a large amount of data is transmitted in the related art.

Fig. 3 is a chart showing the format of a data burst according to a first embodiment of the present invention.

Figs. 4A and 4B are block diagrams showing a

20 transmitting end in an embodiment of the present invention.

Figs. 5A and 5B are block diagrams showing a receiving end in an embodiment of the present invention.

Fig. 6 is a flowchart showing a case where a preamble is being received in the first embodiment of the present

25 invention.

Fig. 7 is a flowchart showing a case where a mid-amble is being received in the first embodiment of the present invention.

Fig. 8 is a chart showing the format of data bursts in a second embodiment of the present invention.

Fig. 9 is a flowchart showing a case where a mid-amble is being received in the second embodiment of the present invention.

Fig. 10 is a chart showing the format of data bursts according to a third embodiment of the present invention.

Fig. 11 is a flowchart showing a case where a preamble is being received in the third embodiment of the present invention.

Fig. 12 is a flowchart showing a case where a mid-amble is being received in the third embodiment of the present invention.

Fig. 13 is a chart showing the format of data bursts according to a fourth embodiment of the present invention.

DESCRIPTION OF THE PREFERRED EMBODIMENTS

Fig. 3 shows the format of a data burst according to a first embodiment of a digital communication method of the present invention. For example, the data burst having a train-type data-burst structure includes a preamble (PR), data (DA) and one or a plurality of what will, hereinafter,

be termed "mid-ambles" (MD). The preamble (PR) includes a synchronization code (SY), a unique word representing the start of received data, and a station-identification code (ID) showing which station the information is addressed to.

5 The mid-amble (MD) includes a synchronization code (SY).

The length of the data (DA) is equal to the maximum duration of data capable of being transmitted with one data burst, which is limited by a time during which synchronous precision or AGC precision in the related art can be

10 maintained.

Figs. 4A, 4B, 5A and 5B show the diagrams of a transmitting end and a receiving end according to the first embodiment of the present invention. As shown in Fig. 4A, a data processor 41 in the transmitting end generates a data

15 burst as shown in Fig. 3 in accordance with a command from an upper layer 40, and sends the data burst as a spread spectrum signal to a transmission line via a high-frequency processor 42. Information data (DA), a station-identification code (ID), a status indicator (ST) (shown in

20 Fig. 10) and so forth are sent as data from the upper layer 40 into the data processor 41 shown in Fig. 4A. A timing generator 41A shown in Fig. 4B generates each predetermined timing involved in constructing the data burst. In accordance with the timing generated by the timing generator

25 41A, an SS modulator 41E outputs the synchronization code

(SY) without modulating it in the synchronization-code (SY) period. Also, in the unique-word (UW) period a selector 41D selects a transmission signal from a signal series "0101..." generated from a UW generator 41C in accordance with the timing generated by the timing generator 41A. The SS modulator 41E performs the spread modulation of the output of the selector 41D, and the modulated output is sent as a spread spectrum signal to the transmission line via the high-frequency processor 42. In the synchronization-code (SY) period the selector 41D operates so that the SS modulator 41E is not supplied with the signals from the upper layer 40 and the UW generator 41C.

At this time the timing generator 41A has control such that the synchronization code (SY) is output from the SS modulator 41E in the SY period, the unique word (UW) is output from the UW generator 41C in the unique-word (UW) period among the other periods, and the station-identification code (ID), the status (ST) and the data (DA) are output from the upper layer 40.

Figs. 6 and 7 show flowcharts of the operation of the receiving end when it has received the above-described data burst. Fig. 6 shows a condition in which the preamble (PR) is being received, while Fig. 7 shows a condition in which the mid-amble (MD) is being received.

When the receiving end, having received the spread-

spectrum data burst, receives the synchronization code (SY) of the preamble (PR) in step S11, the AGC is acquired by a high-frequency processor 43 in steps S12 and S13, and clock synchronization is established by a synchronizer 44A in

5   accordance with the synchronization code in steps S14 and S15. In the successive steps S16 and S17, while the AGC and the clock synchronization are fine-adjusted, the unique word (UW) is detected by a comparator 44E in steps S18 and S19. In an initial condition a selector 44G selects a UW

10 generator 44D. When the comparator 44E detects the unique word (UW), in step S20 a timing generator 44F holds the AGC and the clock synchronization by the high-frequency processor 43 and the synchronizer 44A, and switches the selector 44G to an ID generator 44B in step S21. In step

15 S21, the comparator 44E detects the station-identification code (ID), and when it recognizes that the received data is desired data addressed to the receiving end (i.e., to this particular receiver), the timing generator 44F causes a demodulator 44C to demodulate the data (DA) in step S215.

20   Termination of the data (DA) in step S22 is followed by determination of whether or not the next group of data (DA) is being received in step S31. If the next data (DA) is being received (or has arrived in its entirely), reception of the synchronization code (SY) of the mid-amble (MD)

25 causes the timing generator 44F to perform fine adjustment

of the AGC by the high-frequency processor 43 in step S32, fine phase adjustment of the clock signal by the synchronizer 44A in step S33, and so forth. An upper layer 45 informs the timing generator 44F whether the data has

5   terminated or whether a following mid-amble has been received, in step S31. Since acquisition of the clock synchronization and the AGC has been established in the preamble (PR) period, the clock synchronization in the mid-amble (MD) period is sufficiently achieved by only phase

10   correction, and the initial acquisition of the AGC is not needed. Accordingly, the synchronization code in the mid-amble (MD) period may be shorter than the synchronization code in preamble (PR) period. The high-frequency processor 43 increases the gain to cause the rapid acquisition of the

15   AGC in step S13, and decreases the gain to perform fine adjustment of the AGC in step S16 or S32. After the lapse of a predetermined time, the timing generator 44F holds the AGC and the clock synchronization by the high-frequency processor 43 and the synchronization unit 44 in step S36,

20   and causes an SS demodulator 44C to demodulate the data (DA) in step S215. The receiving end performs the above processes until the train terminates.

In this manner, according to the first embodiment, even when a large amount of digital data is sent, the data can be

25   transmitted without separating it into a plurality of data

bursts. In this case the need for a plurality of conventionally required unique words (UW), station-identification codes (ID) and guard times (GT) is eliminated, and there is no possibility that interrupting

5 bursts from another station occur. Consequently, improving the data throughput itself is realized.

Fig. 8 shows the format of a data burst according to a second embodiment of the present invention.

The data burst according to the second embodiment has a

10 train-type data-burst structure including a preamble (PR), data (DA) and one or a plurality of mid-ambles (MD). The preamble (PR) includes a synchronization code (SY), a unique word (UW) and a station-identification code (ID). Each mid-amble includes a synchronization code (SY) and a unique word

15 (UW).

The structure used with this format is identical to that shown in Figs. 4A, 4B, 5A and 5B.

Fig. 9 shows a flowchart of the operation of the receiving end in handling such a data burst, and in

20 particular, the mid-amble (MD). The operation of the receiving end while receiving the preamble (PR) is identical to that shown in Fig. 6.

In a case where there are successive groups of data (DA), when the synchronization code (SY) of the mid-amble

25 (MD) is received in step S31, the high-frequency processor

43 performs fine adjustment of AGC in step S32 and the synchronizer 44A performs fine phase adjustment of the clock signal and so forth in step S33. Here, since the acquisition of the clock synchronization and the AGC has

5   been established in the preamble (PR) period, the clock synchronization in the mid-amble (MD) period is sufficiently achieved by only phase correction, and the initial acquisition of the AGC is not necessary. Accordingly, the synchronization code in the mid-amble (MD) period may be

10   shorter than the synchronization code in the preamble (PR) period. While the AGC and the clock synchronization are being fine-adjusted in steps S32 and S33, the unique word (UW) is detected by the comparator 44E in steps S34A and S35A. When the mid-amble is received, the timing generator

15   44F switches the selector 44G to the UW generator 44D. [When the comparator 44E detects the unique word (UW), the timing generator 44F holds the AGC and the clock synchronization by the high-frequency processor 43 and the synchronization unit 44 in step S36, and causes the SS demodulator 44C to

20   demodulate the data in step S215.

Therefore, even when a large amount of digital data is sent, the need for a plurality of conventionally required station-identification codes (ID) and guard times (GT) is eliminated, and there is no possibility that interrupting

25   bursts from another station occur. Consequently, improving

the data throughput itself is realized.

Here, although inserting the unique word (UW) in the mid-amble (MD) slightly decreases the total throughput, it is effective in improving data-start detection precision

5   after each mid-amble (MD).

Fig. 10 shows the format of a data burst according to a third embodiment of the present invention. For example, the data burst has a train type data-burst structure including data (DA) as information to be originally sent, a guard time

10   (GT) provided before the start of sending in order to avoid conflict, a preamble (PR), and one or a plurality of mid-ambles (MD) in data transmission. The preamble (PR) includes a synchronization code (SY), a unique word (UW) as a signal series of "0101..." representing the start of

15   received data, station-identification code (ID) showing which station the information is addressed to, and a status indicator (ST) as information about the length of the data, the type of data and the number of data groups included in one train (three groups of data in Fig. 10). The mid-amble

20   (MD) includes a synchronization code (SY) and a unique word (UW).

The structure described in the third embodiment is identical to those shown in Figs. 4A, 4B, 5A and 5B.

The receiving end which received the data burst causes

25   the data processor 44 to establish synchronization by means

of the high-frequency processor 43 shown in Fig. 5A and to demodulate the data. In the data processor 44, a demodulation clock signal with synchronization established in the synchronization code (SY) period is used to perform
5   reverse spread demodulation in the SS demodulator 44C, and the comparator 44E compares the signal series of "0101..." generated from the UW generator 44D and the demodulated data. If the output data of the SS demodulator 44C coincides with the unique word (UW) from the UW generator
10   44D, the timing generator 44F generates each predetermined timing included in the data burst, and sends the information data (DA), the station-identification code (ID), the status (ST) and so forth to the upper layer 45.

Figs. 11 and 12 show flowcharts of the operation of the
15   receiving end when processing the above-described data burst. In an initial condition the timing generator 44F sets the selector 44G to the UW generator 44D. When the data burst is received, in step S52 the AGC is acquired by the high-frequency processor 43 and the clock
20   synchronization is established by the synchronizer 44A in accordance with the synchronization code. Successively, when the unique word (UW) is detected by the comparator 44E in step S53, the timing generator 44F performs setting so that in step S54 the high-frequency processor 43 and the
25   synchronizer 44A hold the AGC and the clock synchronization.

The timing generator 44F also sets the selector 44G to the ID generator 44B. In addition, when the comparator 44E detects that the station-identification code (ID) received in step S57 is addressed to the receiving end (i.e., to this particular receiver), the timing generator 44F recognizes the status input from the SS demodulator 44C in the subsequent step S58. The SS demodulator 44C demodulates the data in step S59, and performs setting so that the demodulated data is output to the upper layer 45 until it detects the end of the data in step S60.

The length and number of the groups of data (DA) are included in the status (ST) recognized in step S58. The timing generator 44F controls the reception sequence, based on them.

In addition, by sending the status (ST) to the upper layer 45, notification of the end of the data and the end of the train may be given from the upper layer 45.

If in step S61 the end of the train is not detected, or when a mid-amble (MD) and data group (DA) are received after reception of a preceding data group (DA), the previously established AGC and synchronization held in steps S52 and S54 are adjusted in step S71 in accordance with the synchronization code (SY) of the mid-amble (MD). When the unique word (UW) of the mid-amble (MD) is detected in step S72, the AGC, the synchronization and so forth are held in

step S73, the process returns to step S59. If the SS demodulator 44C detects the end of the train, the process returns to step S52. Also, if it is found in step S57 that the station-identification code (ID) detected by the SS demodulator 44C is not addressed to the receiving end, the process returns to step S52.

The acquisition of the clock synchronization and the AGC has been established in the preamble period. Thus, the clock synchronization in the mid-amble period is sufficiently achieved by only phase correction, and the need for the initial acquisition of the AGC is eliminated. Accordingly, the synchronization code (SY) in the mid-amble (MD) may be shorter than the synchronization code (SY) in the preamble (PR) period. The high-frequency processor 43 increases the gain in step S52 so that the AGC is rapidly acquired, and in step S71 the gain is reduced to precisely adjust the AGC. The synchronizer 44A increases the acquisition gain in step S52, and decreases it in step S71.

Fig. 13 shows the format of a data burst according to a fourth embodiment of the present invention.

The data burst shown in Fig. 13 includes a preamble (PR), data groups (DA) and one or a plurality of mid-ambles (MD). The preamble (PR) includes a synchronization code (SY), a unique word (UW), and a station-identification code (ID). The data (DA) is multiplexed by code division

multiplexing.   Each mid-amble (MD) includes a

synchronization code (SY).   A CDM communication method,

which is one spread-spectrum communication method used to

improve data throughput, uses N mutually orthogonal codes to

5   perform the frequency-axially spread multiplexing of data,

and sends the multiplexed data.

According to the fourth embodiment, the SS modulator

41E has a structure as shown in Fig. 4B, and performs the

code division multiplexing (CDM) of the data burst by using

10   the spread-spectrum (SS) communication method.

In addition, the SS modulator 44C has a structure as

shown in Fig. 5B, and uses N mutually orthogonal PN codes to

perform the CDM reverse spreading of the code-division-

multiplexed data. (The structures shown in SS modulators 41E

15   and 44C are well known in themselves and need not be

described; nonetheless, some details are noted below)   Here,

as shown in Fig. 10, by using a signal which is not

multiplexed as a code in the preamble and the mid-amble,

power consumed by the preamble and the mid-amble can be

20   increased N times power per data channel, which means that

synchronization establishing and AGC inputting, need not be

greatly affected by a change in the communication quality of

the transmission line.

Other operations of the receiving end may be performed

25   by the processes shown in Figs. 6, 7, 9, 11 and 12.

When a code-synchronous CDM communication method is used in the SS modulator 44E, the spread spectrum modulation and multiplexing (CDM) of the selector 41E output is performed using N mutually orthogonal PN (pseudo-noise)

5   codes. In this case the modulator 41E selects a synchronization code PN_0 from spread codes PN_0 to PN_n, and outputs it as an SY code to the high-frequency processor 42, without performing the code division multiplexing of it.

The code-synchronous CDM communication method is a

10   spread-spectrum communication method used to improve the data throughput, which uses N mutually orthogonal codes to perform the spread multiplexing of data onto a frequency, and sends the multiplexed data. When the data or the like formed by the CDM is received and demodulated, gain and

15   synchronization are held by the receiving end.

The station-identification code (ID), the status (ST) and the unique word (UW) other than information data are transmitted using one spread code.

In the foregoing, although the present invention has

20   been described based on the preferred embodiments thereof, the present invention is not limited to the structures of those embodiments but may be modified within the appended claims.